# **CONTROL I**

ELEN3016

#### **Closed-Loop Control Systems**

(Lecture 8)

# Overview

- First Things First!
- Closed-Loop Control Systems
- Examples
- Tutorial Exercises & Homework
- Next Attraction!

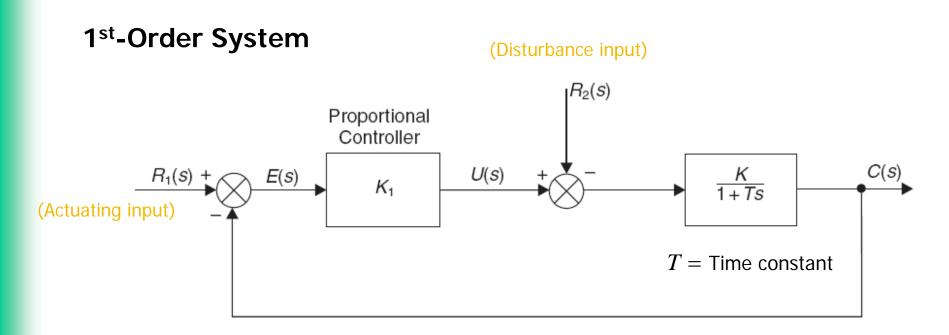


Fig. 4.23 Proportional control of a first-order plant.

$$u(t) = K_1 e(t)$$
  $U(s) = K_1 E(s)$ 

$$C(s) = \frac{\left(\frac{K_1 K}{1 + K_1 K}\right) R_1(s) - \left(\frac{K}{1 + K_1 K}\right) R_2(s)}{\left\{1 + \left(\frac{T}{1 + K_1 K}\right) s\right\}}$$
(4.64)

$$c(t) \begin{vmatrix} c(t) \\ = \left(\frac{K_1 K}{1 + K_1 K}\right) & \text{as } t \to \infty. \\ \begin{pmatrix} Response \text{ to the } \\ Actuating input \end{pmatrix} \\ \begin{pmatrix} K \\ r_2(t) = 0 \end{pmatrix} & \text{as } t \to \infty. \end{aligned}$$

$$c(t) = -\left(\frac{K}{1+K_1K}\right) \quad \text{as } t \to \infty. \quad \begin{array}{l} \text{(Response to the} \\ \text{Disturbance input)} \end{array}$$

Ideally we require that

$$\begin{pmatrix} K_1 K \\ \overline{1 + K_1 K} \end{pmatrix} = 1$$

$$\begin{pmatrix} K \\ \overline{1 + K_1 K} \end{pmatrix} = 0$$

$$(4.65)$$

requiring that  $K_1 K = \infty$  (physically impossible!). Practically all we can do is to set  $K_1 K$  to be a large as is possible, yielding

$$\begin{pmatrix} K_1 K \\ \overline{1 + K_1 K} \end{pmatrix} \approx 1$$

$$\begin{pmatrix} K \\ \overline{1 + K_1 K} \end{pmatrix} \approx 0$$

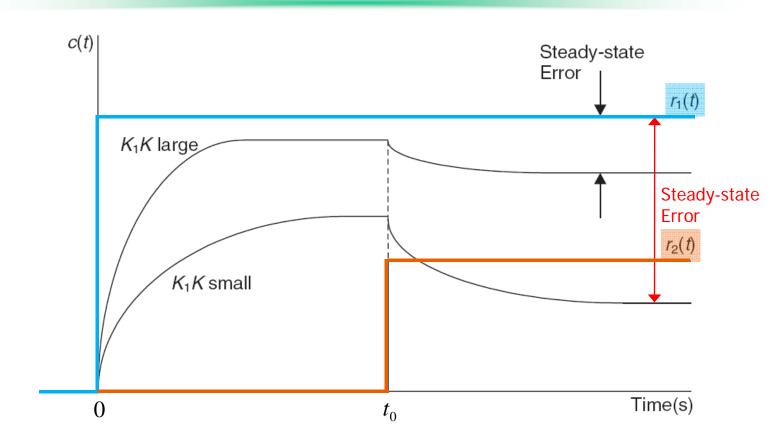


Fig. 4.24 Step response of a first-order plant using proportional control.

Notice: The non-zero steady-state error.

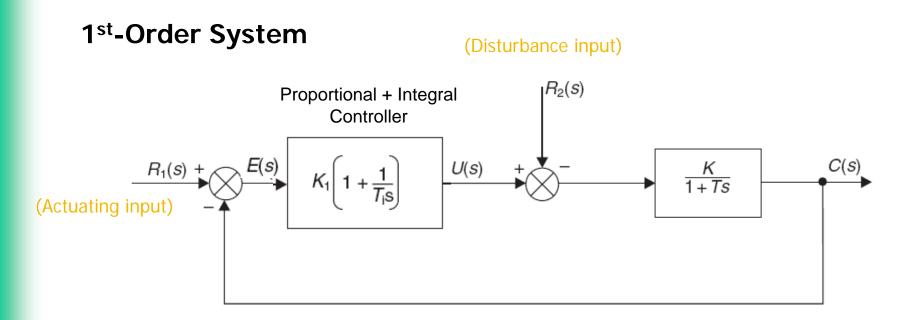


Figure. Proportional + Integral control of the first-order plant.

Open-loop poles: 
$$s = 0$$
,  $s = -\frac{1}{T_i}$ 

$$u(t) = K_1 e(t) + K_2 \int e dt$$
(4.67)

$$U(s) = \left(K_1 + \frac{K_2}{s}\right)E(s)$$
  
=  $K_1\left(1 + \frac{K_2}{K_1s}\right)E(s)$   
=  $K_1\left(1 + \frac{1}{T_1s}\right)E(s)$  (4.68)

Superposition yields

$$C(s) = \frac{(1+T_{i}s)R_{1}(s) - T_{i}sR_{2}(s)/K_{1}}{\left(\frac{T_{i}T}{K_{1}K}\right)s^{2} + T_{i}\left(1 + \frac{1}{K_{1}K}\right)s + 1}$$

$$(4.71)$$

$$\omega_{n}^{-2} \qquad 2\zeta\omega_{n}^{-1}$$

FVT

$$c(t)\Big|_{t\to\infty} = \lim_{s\to0} s C(s)$$
$$= r_1(t)\Big|_{t\to\infty} - 0 \cdot r_2(t)\Big|_{t\to\infty}$$
$$= r_1(t)\Big|_{t\to\infty}$$

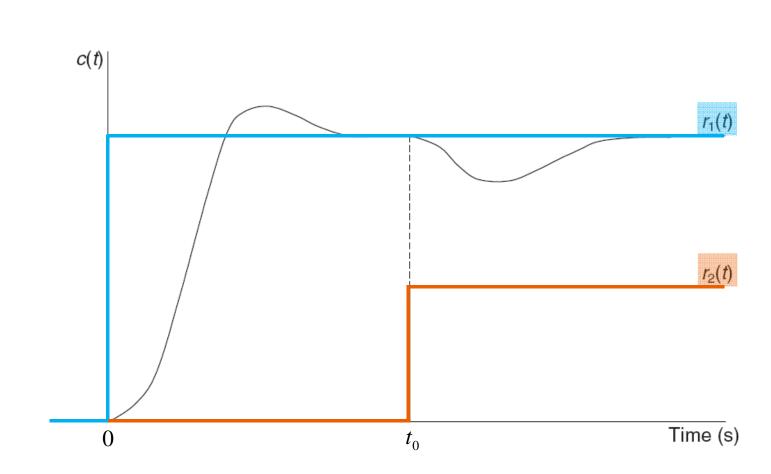


Fig. 4.25 Step response of a first-order plant using PI control.

## **PID** Control

$$u(t) = K_1 e(t) + K_2 \int e dt + K_3 \frac{de}{dt}$$
(4.90)

Taking Laplace transforms, yields

$$U(s) = \left(K_{1} + \frac{K_{2}}{s} + K_{3}s\right)E(s)$$
  
=  $K_{1}\left(1 + \frac{K_{2}}{K_{1}s} + \frac{K_{3}}{K_{1}}s\right)E(s)$   
=  $K_{1}\left(1 + \frac{1}{T_{1}s} + T_{d}s\right)E(s)$  (4.91)  
 $U(s) = \frac{K_{1}(T_{1}T_{d}s^{2} + T_{1}s + 1)}{T_{1}s}E(s)$  (4.92)

# **PID** Control

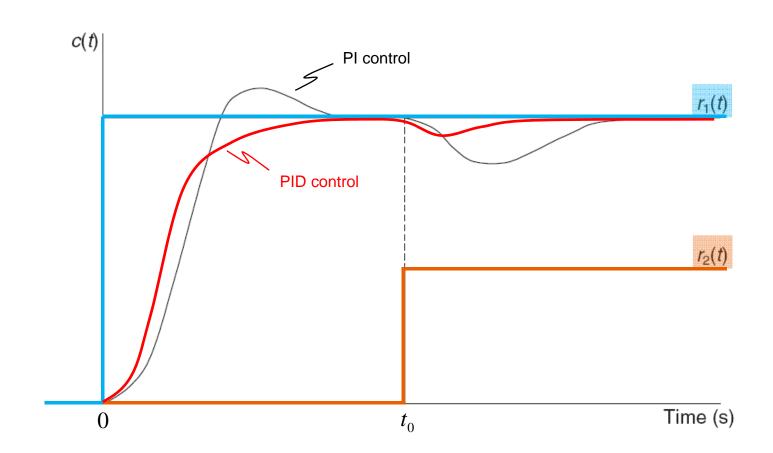
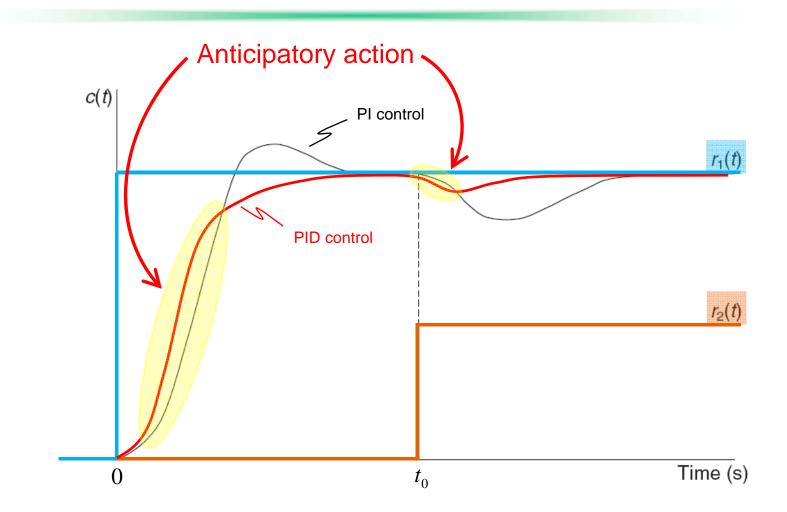


Figure. Proportional + Integral + Derivative control of the first-order plant.

## **PID** Control



**Figure.** Proportional + Integral + Derivative control of the first-order plant.

## **Tutorial Exercises & Homework**

- Tutorial Exercises
  - Example 4.9

- Homework
  - Burns, Example 4.5

## Conclusion

- Closed-Loop P-control of 1<sup>st</sup>-order systems
- Closed-Loop PI-control of 1<sup>st</sup>-order systems
- Closed-Loop PID-control
- Example 4.5 (Self-study!)
- Tutorial Exercises & Homework

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Next Attraction! – Miss It & You'll Miss Out!
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- Ziegler-Nichols Tuning Method
- Case Study

# Thank you!

# **Any Questions?**