### **CONTROL I**

**ELEN3016** 

Classical Design in the Frequency Domain

(Lecture 17)

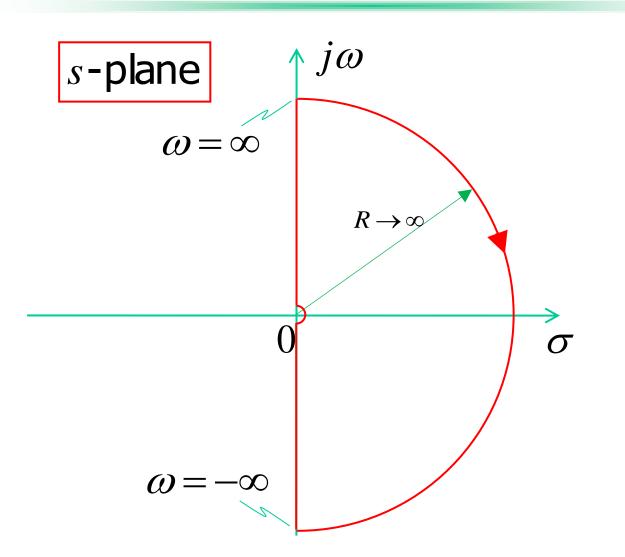
### Overview

- First Things First!
- Nyquist Stability Criterion Examples
- Tutorial Exercises & Homework
- Next Attraction!

# First Things First!

#### • None

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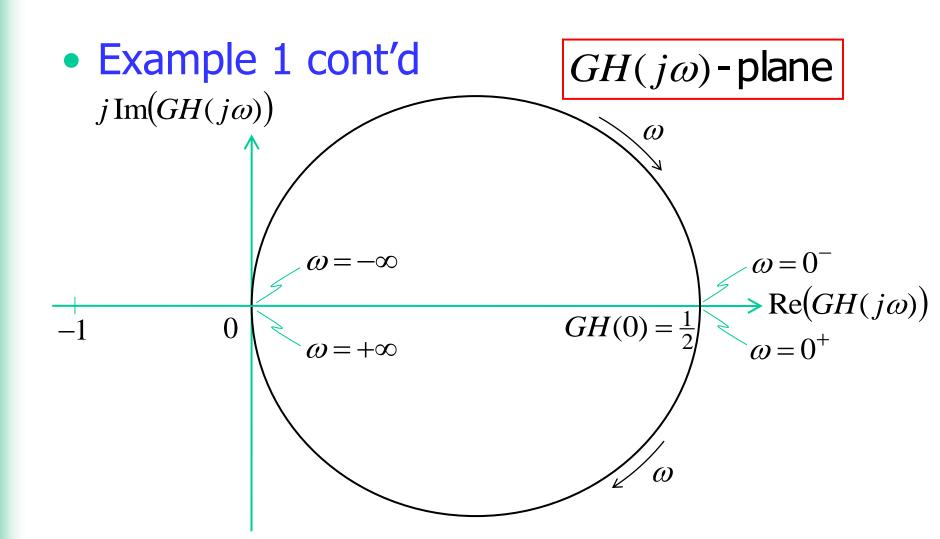


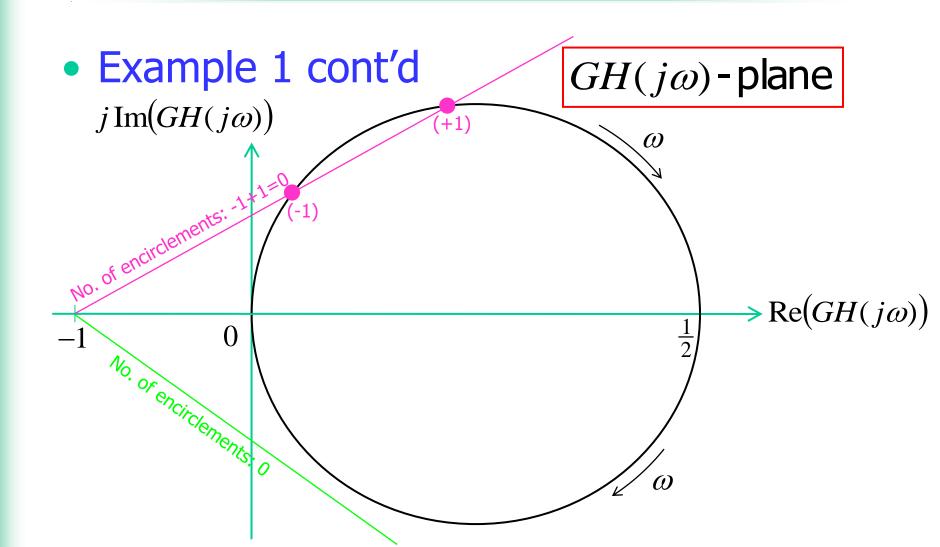
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• Example 1

- First-order system,  $G(s)H(s) = \frac{1}{s+2}$ .

- Open-loop poles: s = -2
- No. of open-loop poles in the RHP: P = 0
- To find N we plot the polar frequency response (*Nyquist plot*.)





- Example 1 cont'd
  - No. of encirclements: N = 0
  - Nyquist criterion:
    No. of closed-loop poles in RHP:

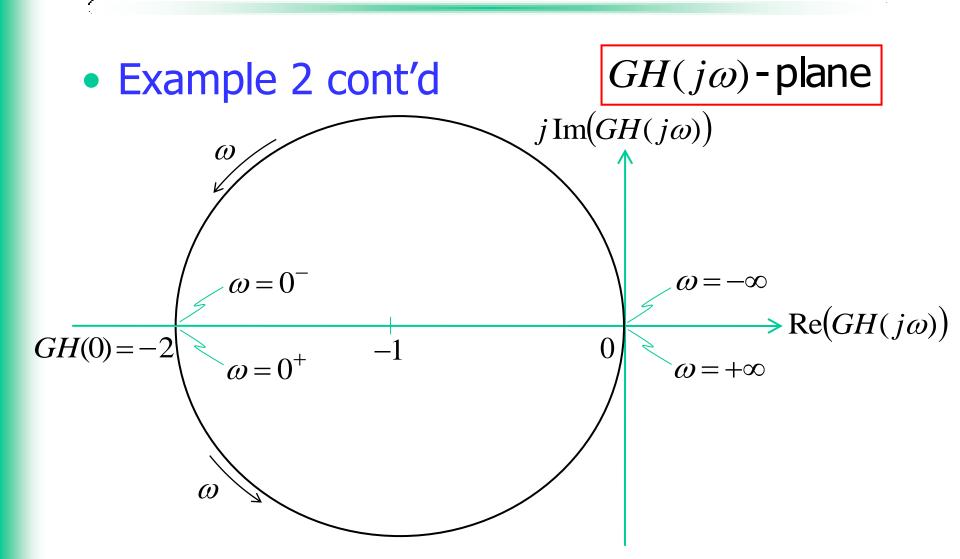
$$Z = N + P = 0 + 0 = 0$$

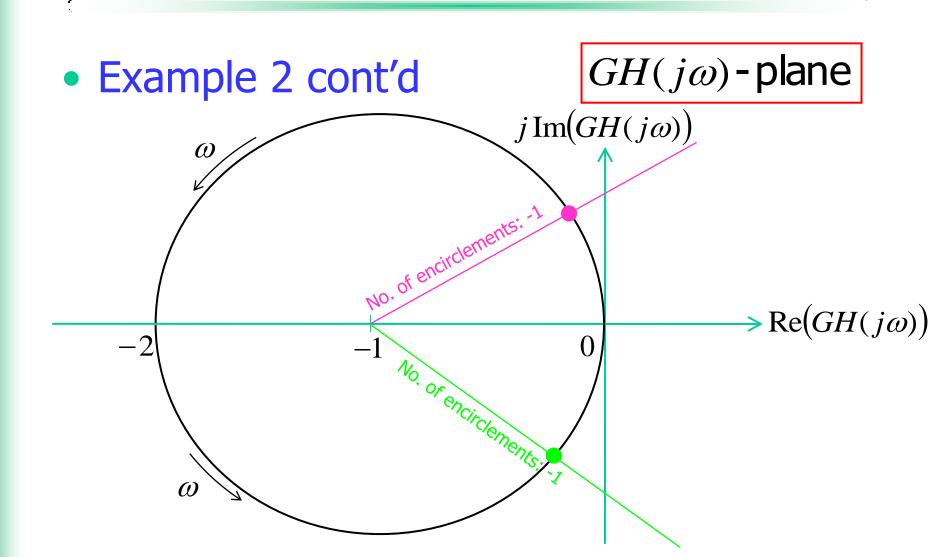
Conclusion: The closed-loop system with characteristic equation 1 + G(s)H(s) is <u>stable</u>.

#### • Example 2

- First-order system,  $G(s)H(s) = \frac{1}{s - \frac{1}{2}}$ .

- Open-loop poles:  $s = \frac{1}{2}$  (unstable!)
- No. of open-loop poles in the RHP: P = 1- To find N we plot the Nyquist plot.





- Example 2 cont'd
  - No. of encirclements: N = -1
  - Nyquist criterion:
    No. of closed-loop poles in RHP:

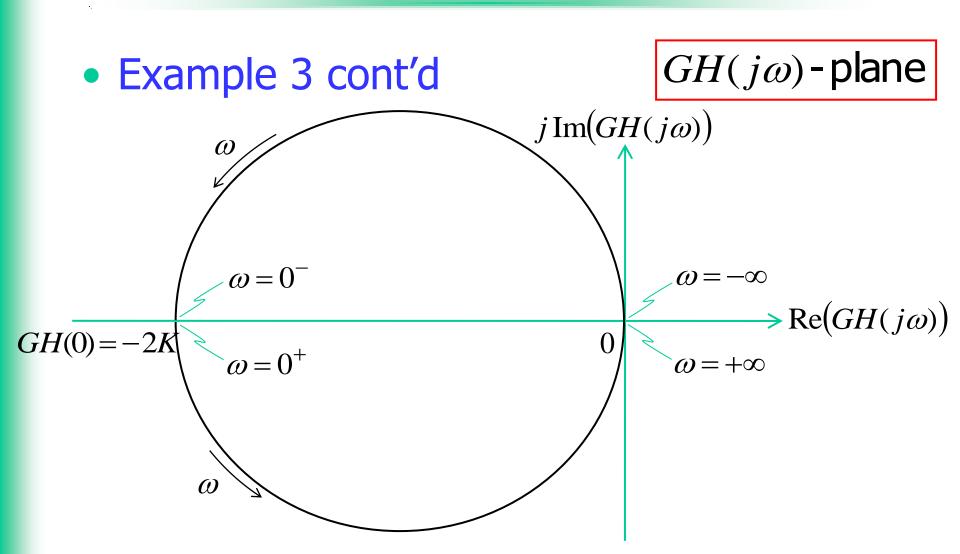
$$Z = N + P = -1 + 1 = 0$$

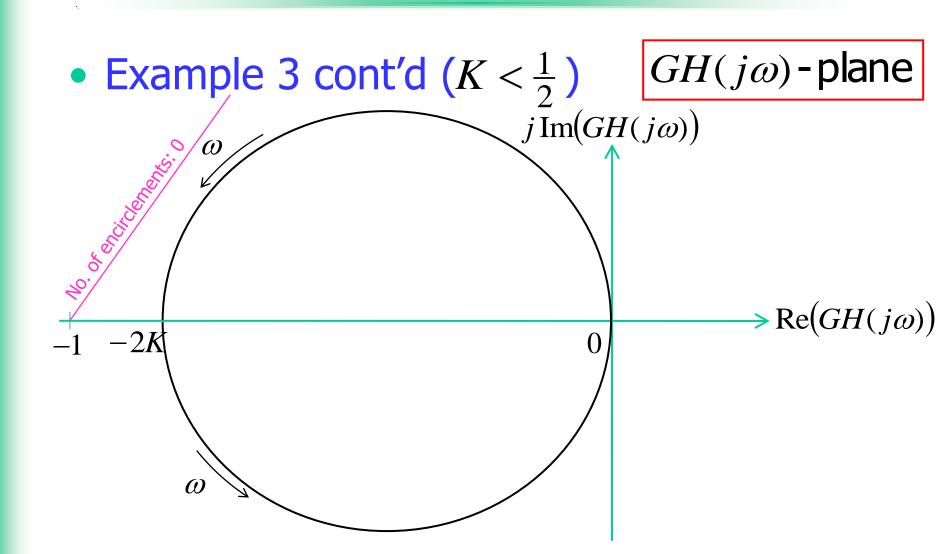
Conclusion: The closed-loop system with characteristic equation 1 + G(s)H(s) is stable.

#### • Example 3

- First-order system,  $G(s)H(s) = \frac{K}{s - \frac{1}{2}}$ , K > 0.

- Open-loop poles:  $s = \frac{1}{2}$  (unstable!)
- No. of open-loop poles in the RHP: P = 1
- To find N we plot the Nyquist plot.

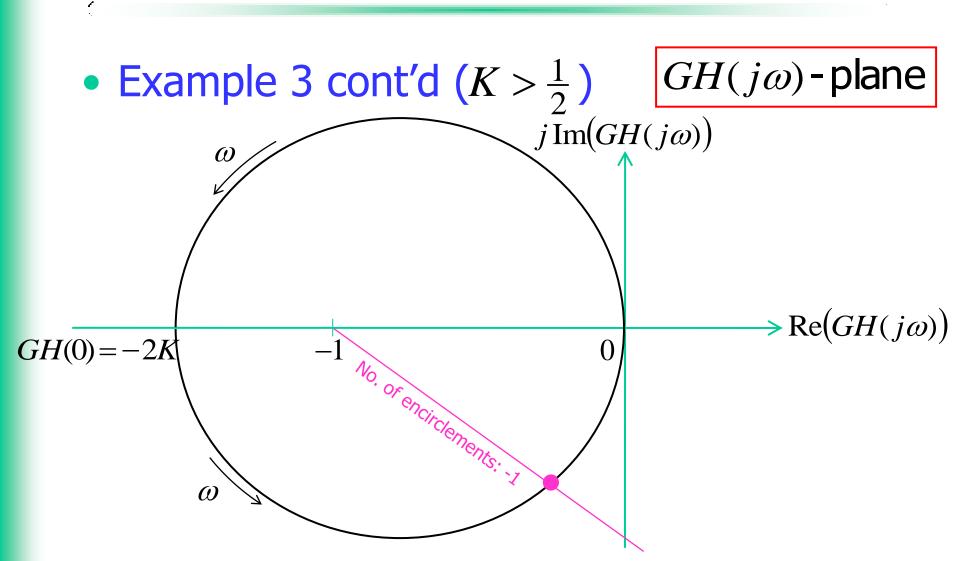




- Example 3 cont'd ( $K < \frac{1}{2}$ )
  - No. of encirclements: N = 0
  - Nyquist criterion:
    No. of closed-loop poles in RHP:

$$Z = N + P = 0 + 1 = 1$$

Conclusion: The closed-loop system with characteristic equation 1 + G(s)H(s) is <u>unstable</u>.



- Example 3 cont'd ( $K > \frac{1}{2}$ )
  - No. of encirclements: N = -1
  - Nyquist criterion:
    No. of closed-loop poles in RHP:

$$Z = N + P = -1 + 1 = 0$$

Conclusion: The closed-loop system with characteristic equation 1 + G(s)H(s) is stable.

### • Example 3 cont'd (K < 0)

- Since *K* is a (multiplicative) factor of the complex number  $G(j\omega)H(j\omega)$ , interpreted as a vector  $G(j\omega)H(j\omega)|_{K<0}$  points in the direction of  $-G(j\omega)H(j\omega)|_{K>0}$ .
- Conclusion: The plots  $G(j\omega)H(j\omega)\Big|_{K<0}$  and  $G(j\omega)H(j\omega)\Big|_{K>0}$  are symmetric about the origin.

### **Tutorial Exercises & Homework**

• Tutorial Exercises

To be announced at the beginning of the tut session.

#### • Homework

- Study all relevant sections in Burns.
- Burns, Example 6.4,
- Burns, Stability on the Bode diagram (pp. 170-171)

### Conclusion

- Introductory Examples
- Burns, Example 6.4 (Self-study!)
- Burns, Stability on the Bode diagram (pp. 170-171) (Self-study!)
- Burns, Section 6.5 (Omit)
- Tutorial Exercises & Homework

Next Attraction! – Miss It & You'll Miss Out!

### Steady-State Error Analysis (Interlude)

# Thank you! Any Questions?