CONTROL I

ELEN3016

1

Stability of Dynamical Systems

(Lecture 11)

Overview

- First Things First!
- Introductory Examples
- Review of 2nd-Order Systems
- Routh-Hurwitz Stability Criterion
- Tutorial Exercises & Homework
- Next Attraction!

First Things First!

Lab Related Matters

– Student assistants:

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- DLab booked: Mondays & Tuesdays, 14:00 17:00
- Official Deadline for to be decided
- Informal milestone dates?

First Things First!

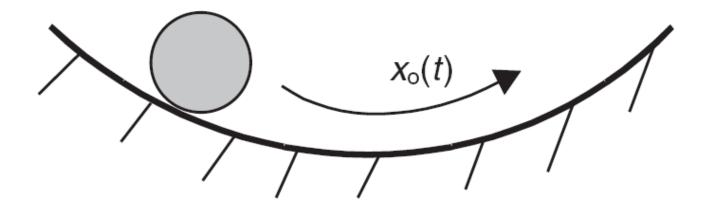
Test

- After the study break ... 😳
 - 30th September, 08:00 10:00 in FNBBA.
- Will cover Chapters 1 to 5 of Burns.
- All problems discussed in class, self-study, homework, all tutorial problems and (small) variations on these themes are important for the Test.

Introductory Example 1

• *Stable* time response

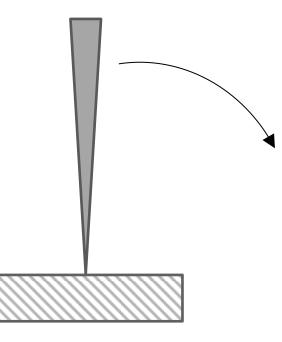
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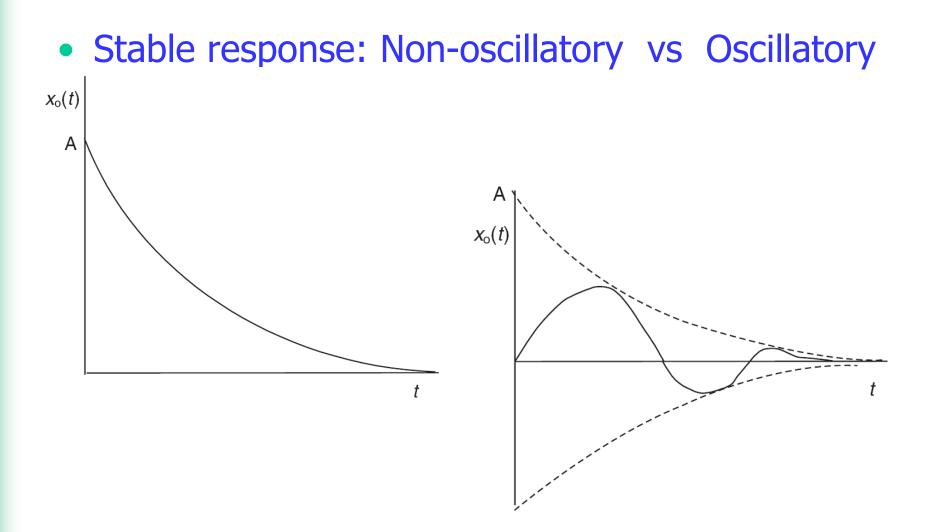
Introductory Example 2

• Example of *unstable* time response

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Types of Stable Time Response



Types of Unstable Time Response

Unstable response: Non-oscillatory vs Oscillatory $x_{\rm o}(t)$ $x_{o}(t)$ A A t t

Characteristic equation of a 2nd-order system

$$as^2 + bs + c = 0 (5.5)$$

 $\begin{cases} \operatorname{Re}(s_1), \operatorname{Re}(s_2) < 0\\ b > \sqrt{b^2 - 4ac} \end{cases}$

• Roots

$$s_1, s_2 = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a} \tag{5.6}$$

$$a, b, c > 0 \implies$$

Thatis, all of the same sign!

- Overdamped 2nd-order system $(b^2 4ac > 0)$ $s_1 = -\sigma_1$ $s_2 = -\sigma_2$ (5.7)
- Critically damped 2nd-order system ($b^2 4ac = 0$) $s_1 = s_2 = -\sigma$ (5.8)
- Underdamped 2nd-order system ($b^2 4ac < 0$)

$$s_1, s_2 = -\sigma \pm j\omega \tag{5.9}$$

- 2nd-order system with a, c > 0, b < 0 implies
 a. -4ac < 0 ⇒ b² 4ac < b² implying either
 i. complex conjugate pole pair if b²-4ac < 0 or
 ii. two real poles of the same sign if 0 < b²-4ac.
 - **b.** $-\frac{b}{2a} > 0$ i.e. at least one unstable pole.

Combining a.) & b.) we conclude that both poles (real or complex) are in the RHP giving an <u>unstable</u> system.

- 2nd-order system with b, c > 0, a < 0 implies a. $-4ac > 0 \Rightarrow b^2 - 4ac > b^2$ from which we have two real poles of opposite signs since $0 < b < \sqrt{b^2 - 4ac}$.
 - **b.** $-\frac{b}{2a} > 0$ i.e. at least one unstable pole.

Combining a.) & b.) ensures a single unstable pole in the RHP giving an <u>unstable</u> system.

- 2nd-order system with a, b > 0, c < 0 implies a. $-4ac > 0 \Rightarrow b^2 - 4ac > b^2$ from which we have two real poles of opposite signs since $0 < b < \sqrt{b^2 - 4ac}$.
 - **b.** $-\frac{b}{2a} < 0$ i.e. at least one stable pole.

Combining a.) & b.) ensures a single unstable pole in the RHP giving an <u>unstable</u> system.

Stability of a LTI System

• Characteristic equation of an *n*th-order system

$$a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0 = 0$$

- Necessary <u>but not</u> sufficient condition for a system to be stable:
 - The coefficients of the characteristic equation must all have the <u>same</u> sign and that none are zero.

Stability of a LTI System

• Set-theoretic depiction of all LTI *n*th order systems

Set of all LTI *n* th order systems

Set of all LTI *n* th order systems with all coefficients positive

Set of all stable LTI *n* th order systems

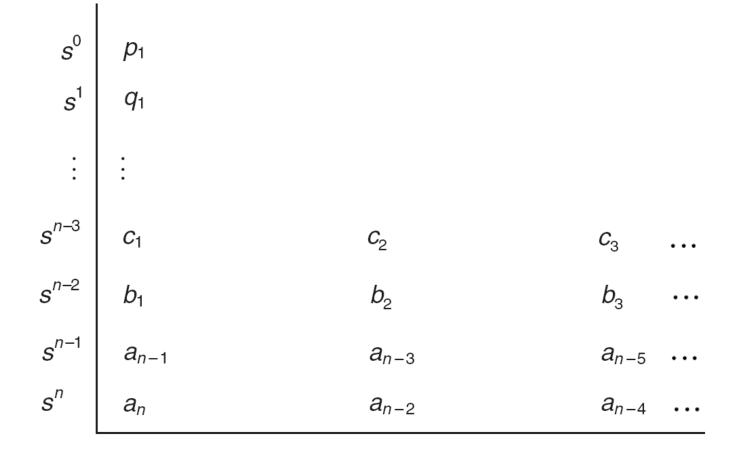
Comment on the closedness and openness of these sets.

• Characteristic equation of an *n*th-order system

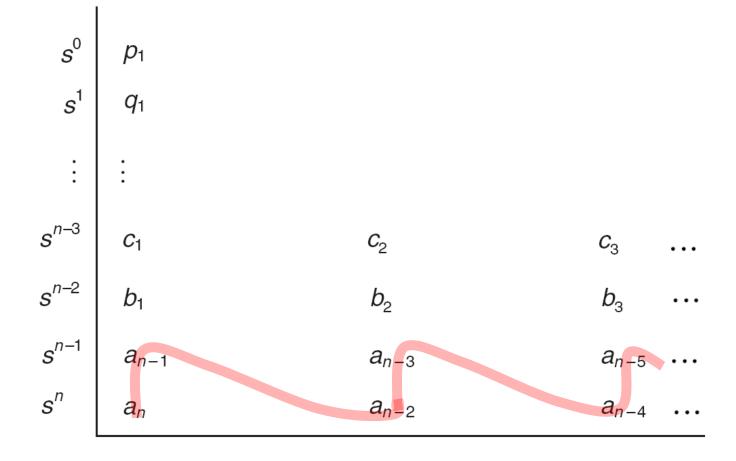
 $a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0 = 0$ (5.11)

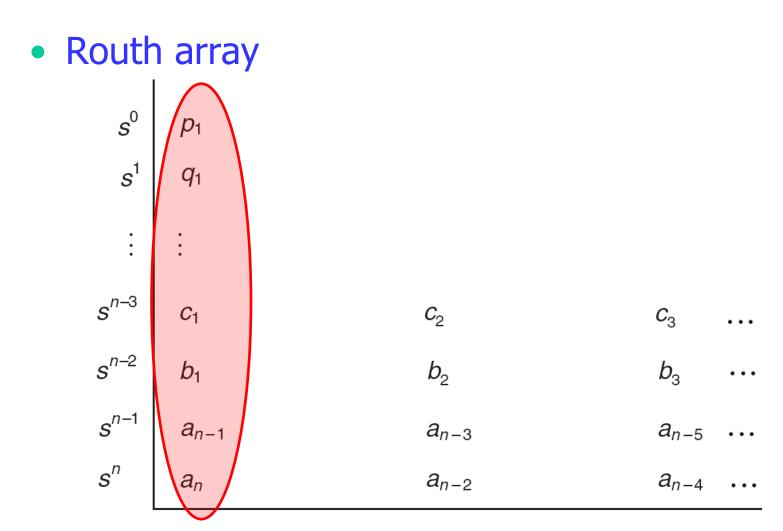
- Necessary <u>and</u> sufficient condition for a system to be stable:
 - All the *Hurwitz determinants* of the characteristic polynomial are positive or, equivalently,
 - All coefficients in the first column of the *Routh array* have the same sign.

• Routh array



• Routh array





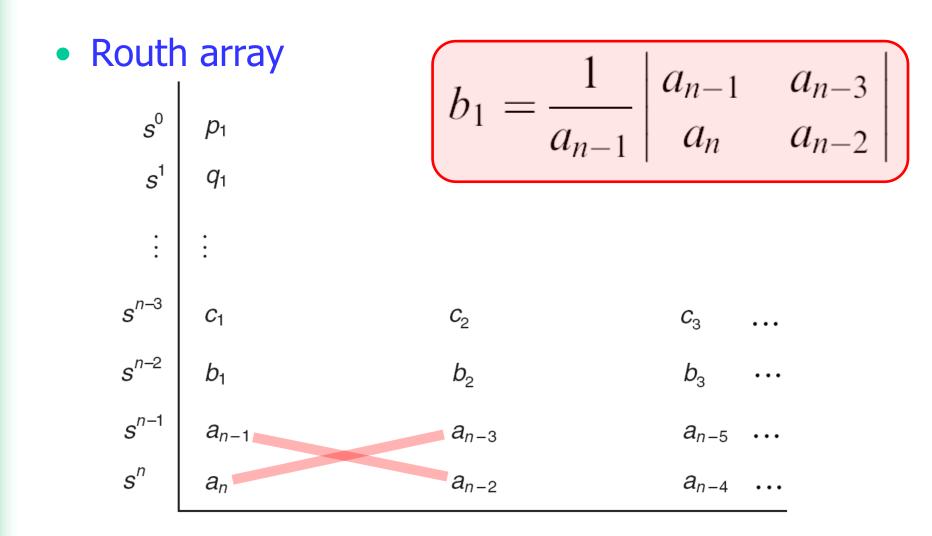
Developing the Routh array's third row:

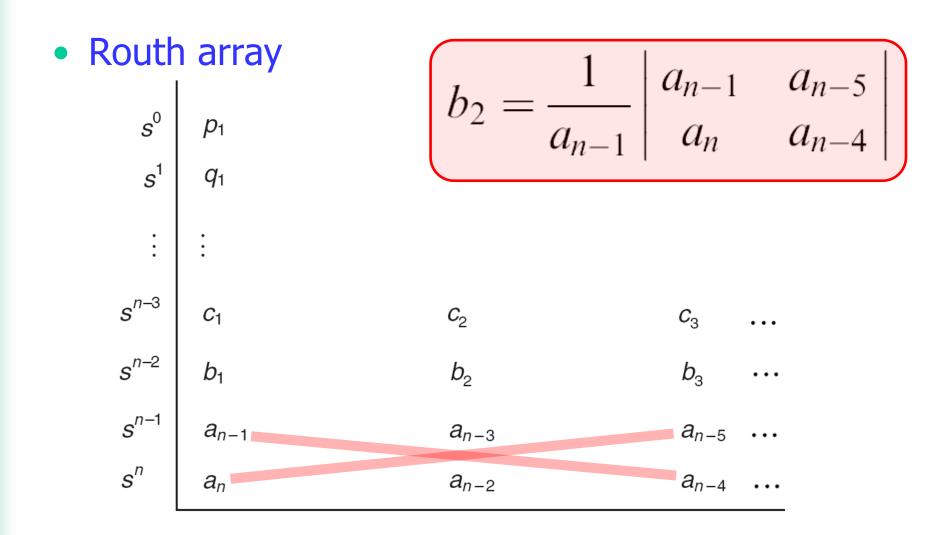
$$b_1 = \frac{1}{a_{n-1}} \begin{vmatrix} a_{n-1} & a_{n-3} \\ a_n & a_{n-2} \end{vmatrix}$$

$$b_2 = \frac{1}{a_{n-1}} \begin{vmatrix} a_{n-1} & a_{n-5} \\ a_n & a_{n-4} \end{vmatrix}$$

etc.

Continue until the first zero appears. Then move to next row.



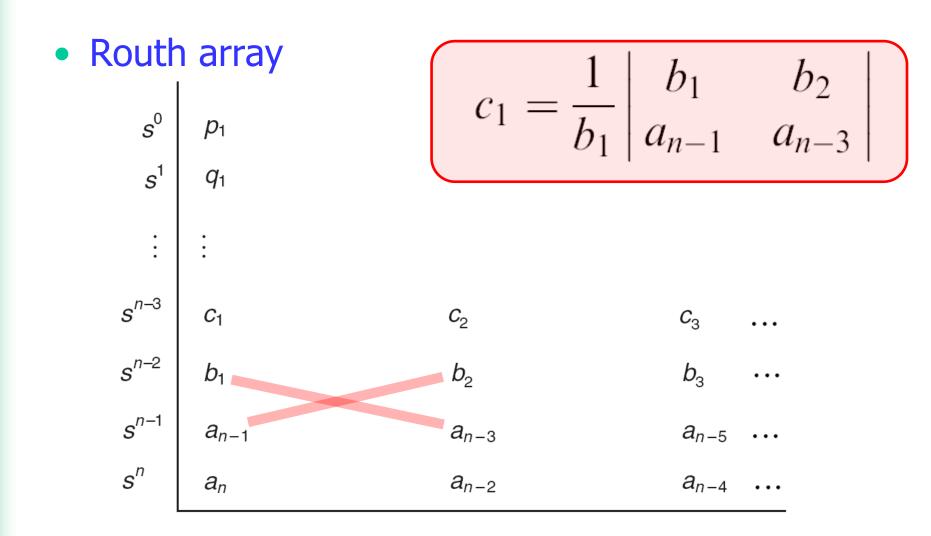


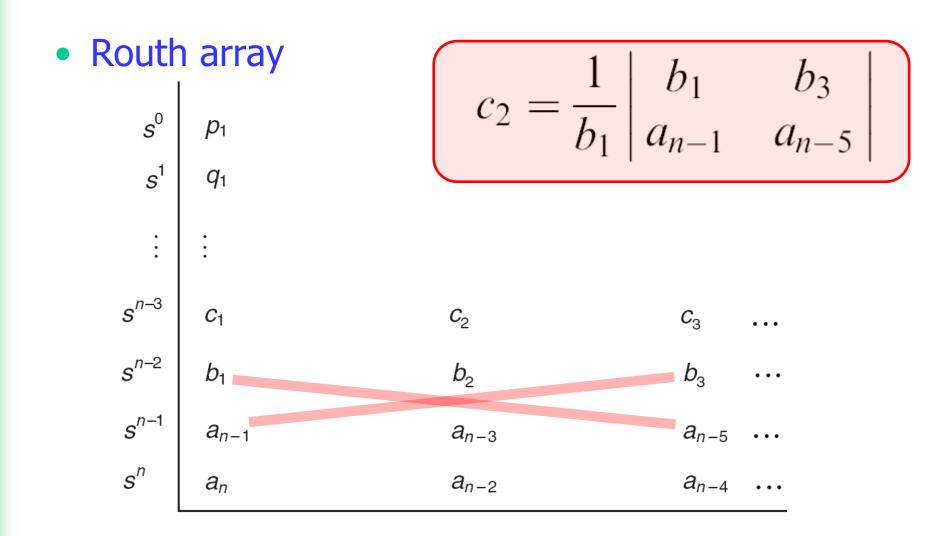
Developing the Routh array's fourth row:

$$c_{1} = \frac{1}{b_{1}} \begin{vmatrix} b_{1} & b_{2} \\ a_{n-1} & a_{n-3} \end{vmatrix}$$
$$c_{2} = \frac{1}{b_{1}} \begin{vmatrix} b_{1} & b_{3} \\ a_{n-1} & a_{n-5} \end{vmatrix}$$

etc.

 Continue until the first zero appears. Then move to next row. Terminate with a complete row of zeros.





• Example 5.1 $s^4 + 2s^3 + s^2 + 4s + 2 = 0$

Routh array:

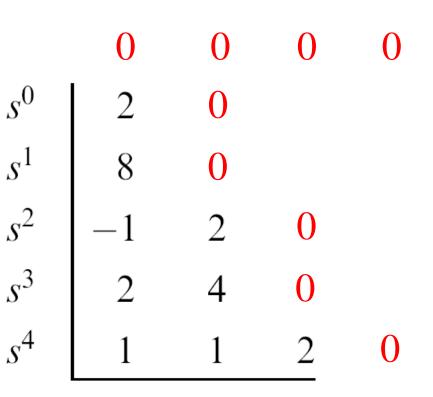
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• Example 5.1 (cont'd)

- Below shown the zeros that guide the calculations.

Routh array:

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• Example 5.1 (cont'd)

- Below shown the zeros that guide the calculations.

 s^0

 s^1

 s^2

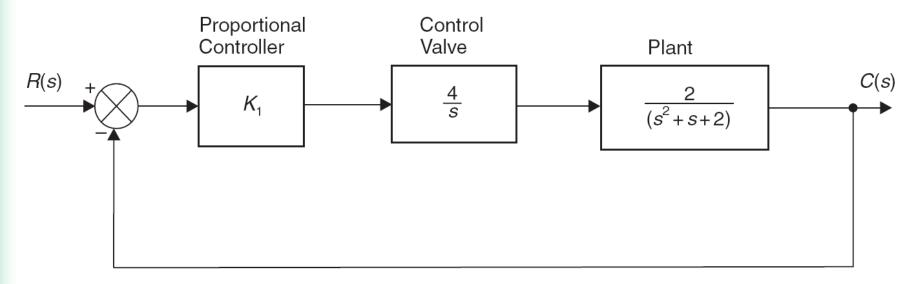
 s^3

 s^4

Routh array:

- 2 sign changes
- 2 poles in RHP

• Example 5.2



Find the minimum proportional gain for which the system is marginally stable (i.e. "only just unstable").

• Example 5.2 (cont'd)

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Characteristic equation: $(K = 8K_1)$

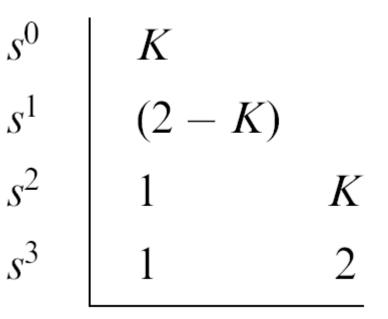
$$1 + G(s)H(s) = 1 + \frac{K}{s(s^2 + s + 2)} = 0$$

$$s^3 + s^2 + 2s + K = 0 \tag{5.30}$$

• Example 5.2 (cont'd)

Routh array:

1



• Example 5.2 (cont'd)

1

Routh array: Suppose $K = \varepsilon$, $0 < \varepsilon \ll 1$ - 0 sign changes $\begin{vmatrix} s^0 & \varepsilon \\ s^1 & 2 \\ s^2 & 1 & \varepsilon \\ s^3 & 1 & 2 \end{vmatrix}$

• Example 5.2 (cont'd)

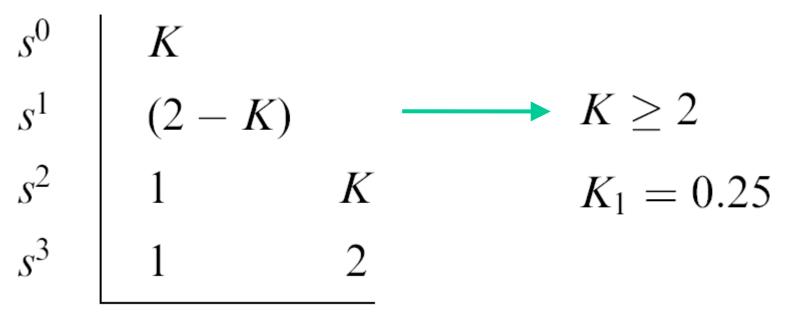
Routh array:Suppose $K = 2 - \varepsilon$, $0 < \varepsilon \ll 1$ - 0 sign changes s^0 2- 0 poles in RHP s^1 ε s^2 12 s^3 12

• Example 5.2 (cont'd)

Routh array:Suppose $K = 2 + \varepsilon$, $0 < \varepsilon \ll 1$ - 2 sign changes s^0 2- 2 poles in RHP s^1 $-\varepsilon$ s^2 12 s^3 12

• Example 5.2 (cont'd)

Stability requires no sign changes in 1st column.
 Routh array:



Example 5.2 (cont'd)

Study the remaining part of Example 5.2 as well as Section 5.2.2 on *special cases* of the Routh array.

Tutorial Exercises & Homework

- Tutorial Exercises
 - Burns, Examples 5.12 and 5.13

- Homework
 - Burns, Example 5.2 and Sec. 5.2.2

Conclusion

- First Things First!
- Introductory Examples
- Review of 2nd-Order Systems' Stability
- Routh-Hurwitz Stability Criterion
- Burns, Sec 5.2.2 (Self-study!)
- Tutorial Exercises & Homework

Next Attraction! – Miss It & You'll Miss Out!

The Root Locus Technique (Burns, Chapter 5)

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Thank you! Any Questions?